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FORM PTO-1390
(REV. 5-93)

U.S. DEPARTMENT OF COMMERCE PATENT AND TRADEMARK OFFICE

TRANSMITTAL LETTER TO THE UNITED STATES DESIGNATED/ELECTED OFFICE (DO/EO/US) CONCERNING A FILING UNDER 35 U.S.C. 371

ATTORNEY'S DOCKET NUMBER
518-1012

U S APPLICATION NO (If known, see 37 C F.R 1 5)
09/673002

INTERNATIONAL APPLICATION NO	INTERNATIONAL FILING DATE	PRIORITY DATE CLAIMED
PCT/FR99/00835	April 9, 1999	April 10, 1998

TITLE OF INVENTION
Method For Fine Synchronization On A Signal Received From A Transmission Channel

APPLICANT(S) FOR DO/EO/US
Jean-Louis Dornstetter and Nidham Ben Rached

Applicant herewith submits to the United States Designated/Elected Office (DO/EO/US) the following items and other information:

1. X This is a **FIRST** submission of items concerning a filing under 35 U.S.C. 371.

2. _ This is a **SECOND** or **SUBSEQUENT** submission of items concerning a filing under 35 U.S.C. 371.

3. _ This express request to begin national examination procedures (35 U.S.C. 371(f) at any time rather than delay examination until the expiration of the applicable time limit set in 37 U.S.C. 371(b) and PCT Articles 22 and 39(1).

4. X A proper Demand for International Preliminary Examination was made by the 19th month from the earliest claimed priority date.

5. X A copy of the International Application as filed (35 U.S.C. 371(c)(2))
   a. _ is transmitted herewith (required only if not transmitted by the International Bureau).
   b. X has been transmitted by the International Bureau.
   c. _ is not required, as the application was filed in the United States Receiving Office (RO/US).

6. _ A translation of the International Application into English (35 U.S.C. 371(c)(2)).

7. X Amendments to the claims of the International Application under PCT Article 19 (35 U.S.C. 371(c)(3))
   a. _ are transmitted herewith (required only if not transmitted by the International Bureau).
   b. _ have been transmitted by the International Bureau.
   c. _ have not been made; however, the time limit for making such amendments has NOT expired.
   d. X have not been made and will not be made.

8. _ A translation of the amendments to the claims under PCT Article 19 (35 U.S.C. 371(c)(3)).

9. _ An oath or declaration of the inventor(s) (35 U.S.C. 371(c)(4)).

10. _ A translation of the annexes to the International Preliminary Examination Report under PCT Article 36 (35 U.S.C. 371(c)(5)).

Items 11. to 16. below concern other document(s) or information included:

11. _ An Information Disclosure Statement under 37 CFR 1.97 and 1.98.

12. _ An assignment document for recording A separate cover sheet in compliance with 37 CFR 3.28 and 3.31 is included.

13. _ A FIRST preliminary amendment.
    _ A SECOND or SUBSEQUENT preliminary amendment.

14. _ A substitute specification.

15. _ A change of power of attorney and/or address letter.

16. _ Other items or information:

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U.S. APPLICATION NO (if known, see 37 CFR 1.5) 09/673002	INTERNATIONAL APPLICATION NO PCT/FR99/00835	ATTORNEY'S DOCKET NUMBER 518-1012

	CALCULATIONS	PTO USE ONLY
17. X The following fees are submitted: **Basic National Fee** 37 CFR 1.492(a)(1)-(5)): Search Report has been prepared by the EPO or JPO . . . . $840.00 International preliminary examination fee paid to USPTO (37 CFR 1.482) . $670.00 No international preliminary examination fee paid to USPTO (37 CFR 1.482) but international search fee paid to USPTO (37 CFR 1.445(a)(2)) . . . . $760.00 Neither international preliminary examination fee (37 CFR 1.482) nor international search fee (37 CFR 1.445(a)(2)) paid to USPTO . . . . $970.00 International preliminary examination fee paid to USPTO (37 CFR 1.482) and all claims satisfied provisions of PCT Article 33(2)-(4) $96.00		
ENTER APPROPRIATE BASIC FEE AMOUNT =	$860	
Surcharge of **$130.00** for furnishing the oath or declaration later than _ 20 _ 30 months from the earliest claimed priority date (37 CFR 1.492(e)).		

Claims	Number Filed	Number Extra	Rate	CALCULATIONS	PTO USE ONLY
Total Claims	10 - 20 =	0	X $18.00	0	
Independent Claims	1 - 3 =	0	X $78.00	0	
Multiple dependent claim(s) (if applicable)			+ $260.00	270	

	CALCULATIONS	PTO USE ONLY
TOTAL OF ABOVE CALCULATIONS =	1130	
Reduction by 1/2 for filing by small entity, if applicable. Verified Small Entity Statement must also be filed. (Note 37 CFR 1.9, 1.27, 1.28).		
SUBTOTAL =	1130	
Processing fee of **$130.00** for furnishing the English translation later than _ 20 _ 30 months from the earliest claimed priority date (37 CFR 1.492(f)).		
TOTAL NATIONAL FEE =	1130	
Fee for recording the enclosed assignment (37 CFR 1.21(h)). The assignment must be accompanied by an appropriate cover sheet (37 CFR 3.28, 3.31). **$40.00** per property +		
TOTAL FEES ENCLOSED =	1130	
	Amount to be refunded:	
	Charged	

a. X A check in the amount of $ 1130 to cover the above fees is enclosed.

b. _ Please charge my Deposit Account No. 12-0913 in the amount of $__ to cover the above fees. A duplicate copy of this sheet is enclosed.

c. X The Commissioner is hereby authorized to charge any additional fees which may be required, or credit any overpayment to Deposit Account No. 12-0913 . A duplicate copy of this sheet is enclosed.

NOTE: Where an appropriate lime limit under 37 CFR 1.494 or 1.495 has not been met, a petition to revive (37 CFR 1.137(a) or (b)) must be filed and granted to restore the application to pending status.

SEND ALL CORRESPONDENCE TO:
William M. Lee, Jr.
Lee, Mann, Smith, McWilliams, Sweeney & Ohlson
P.O. Box 2786
Chicago, Illinois 60690-2786
(312) 368-1300

William M. Lee
SIGNATURE

William M. Lee, Jr.
NAME

26,935
REGISTRATION NUMBER

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09/673002
10 Rec'd PCT/PTO 11 DEC 2000

METHOD OF FINE SYNCHRONIZATION TO A SIGNAL RECEIVED FROM A TRANSMISSION CHANNEL

The present invention relates to a method of fine synchronization to a receive signal corresponding to a reference signal transmitted in a transmission channel.

In a transmission system, in particular a radio transmission system, a transmitter transmits a reference signal to a receiver in a transmission channel. One of the first operations that the receiver must perform is to synchronize to the receive signal. This problem is well-known to the skilled person and it is therefore unnecessary to remind the reader here of the various techniques that have been used to obtain such synchronization.

If the signals are digital signals, it is usual to evaluate the synchronization error by means of a time unit which is the time difference between two successive bits of a signal and which is referred to as the bit period. It appears that the available prior art solutions cannot achieve synchronization to an accuracy much better than one bit period.

This accuracy can prove insufficient in some cases. This is because synchronization gives the transit time of the signal in the transmission channel, in other words the transmission time between the transmitter and the receiver. This data is important in a duplex radiocommunications system in which a base station communicates with a terminal because the base station and the terminal have respective transceivers and the terminal must operate in such a way that the signal it transmits arrives at a precise time with reference to the clock of the base station. To achieve this, it is naturally necessary for the terminal to know the time taken for the signal to reach the base station.

The transmission time is also a direct reflection of the distance between the transmitter and the receiver. Clearly, the accuracy of this distance is of fundamental

importance when it is a question of locating the terminal by identifying its position relative to one or more base stations. The general problem of locating a terminal is a very current concern because of its applications, which include cell handover strategies in cellular networks, for example. The security field should also be mentioned, whether in connection with identifying the geographical location of the source of an emergency call or the position of a stolen vehicle equipped with the terminal.

An object of the present invention is a synchronization method whose accuracy is much better than one bit period.

In accordance with the invention, the method of fine synchronization to a receive signal corresponding to a reference signal transmitted in a transmission channel includes the following steps:

- selecting a source signal producing a characterization signal after it has passed through said transmission channel,
- establishing a characterization matrix for estimating the covariance of the characterization signal,
- identifying dominant eigenvalues which are the highest eigenvalues of the characterization matrix,
- calculating the correlation function of the source signal with the sum of the eigenvectors associated with the dominant eigenvalues, and
- searching for the first maximum of the correlation function.

If the time increment adopted for calculating the correlation function is made sufficiently small, and in all cases much less than one bit period, the above method achieves very good accuracy.

A first option is for the number of dominant eigenvalues to be predetermined. This number typically represents 20% to 30% of the dimension of the characterization matrix.

A second option is for the ratio of the sum of the dominant eigenvalues to the sum of the all the eigenvalues to be greater than or equal to a predetermined percentage. In this case the percentage adopted is often greater than 90%, for example 95%.

A third option is for the method also to include a step of estimating the additive noise in the transmission channel, the dominant eigenvalues being such that their sum is less than or equal to the sum of all the eigenvalues less the additive noise.

Also, the additive noise is estimated by normalizing the instantaneous noise, which is evaluated from the receive signal, the reference signal and an estimate of the impulse response of the transmission channel.

The expression for the instantaneous noise is advantageously: $N_0 = S - A.X$, where A is the transmission matrix associated with the reference signal.

Whichever of the above options is adopted, the characterization matrix results from a smoothing operation.

In a preferred embodiment, the characterization signal is an estimate of the impulse response of the transmission channel.

The characterization signal can instead be the receive signal.

The present invention emerges in more detail from the following description of proposed embodiments of the invention, which is given by way of illustrative example and with reference to the accompanying figures, in which:

- Figure 1 shows a first variant of the invention, and
- Figure 2 shows a second variant.

Elements common to both figures are allocated the same reference numbers.

The receiver has already achieved coarse synchronization to the receive signal, to an accuracy of the order of one bit period, using any of the available prior art solutions.

The receive signal corresponds to a reference signal produced by the transmitter and known to the receiver. The reference signal can be known *a priori*, i.e. it can be a training sequence made up of identified symbols. It can also be determined *a posteriori* using techniques referred to generically as blind probing. In this case, during the synchronization procedure, the receiver regenerates from the receive signal the series of symbols forming the reference signal.

It is first necessary to characterize transmission between the transmitter and the receiver. To this end, a source signal produced by the transmitter is selected which yields a characterization signal in the receiver after transmission over the channel.

Of course, if the source signal is the reference signal, the characterization signal is the receive signal itself. This is not always the optimum solution, however, in terms of the complexity and performance of the method of the invention.

Another solution is to retain a modulated pulse as the source signal, in which case the characterization signal is the impulse response of the transmission channel.

For example, the GSM digital cellular mobile telephone system uses a training sequence made up of 26 symbols, the impulse response being generally estimated with five coefficients since it is accepted that the dispersion of the channel is equal to 4.

In this case, the receive signal has a maximum dimension of 22, which is significantly larger than that of the impulse response.

Two embodiments of the invention are therefore examined in succession and with reference to Figure 1, starting with the situation in which the source signal is a modulated pulse.

Estimating the impulse response does not give rise to any problems in itself because many methods are

available for this, for example the least squares method, which is described in particular in patent applications FR 2 696 604 and EP 0 564 849. Briefly, that technique uses a measurement matrix A constructed from the training sequence TS of length n. The matrix A has (n-d) rows and (d+1) columns, where d represents the dispersion of the channel. The element in the ith row and the jth column is the (d+i-j)th symbol of the training sequence. a_i denotes the ith symbol of a sequence TS of 26 symbols:

$$A = \begin{matrix} a_4 & a_3 & a_2 & a_1 & a_0 \\ a_5 & a_4 & a_3 & a_2 & a_1 \\ a_6 & a_5 & a_4 & a_3 & a_2 \\ a_7 & \ldots & & & \\ \ldots & & & & \\ \ldots & & & & \\ a_{25} & \ldots & & & a_{21} \end{matrix}$$

The training sequence is chosen so that the matrix A^tA can be inverted, where the operator $.^t$ represents transposition.

In the receive signal S, the first four symbols s_0 to s_3 are ignored because, given that the dispersion of the channel is 4, they also depend on the unknown symbol transmitted before the training sequence. The receive signal is defined hereinafter as a vector S having for its components the received symbols s_4, s_5, s_6, ..., s_{25}.

The estimated impulse response X therefore takes the form:

$$X = (A^tA)^{-1}A^t.S$$

The next step of the method of the invention is to establish a statistic of this impulse response, where "statistic" means a data set reflecting the average value of the impulse response over an analysis period.

A matrix is therefore constructed for smoothing the various estimates X obtained during the analysis period to obtain an estimate of the covariance associated with

the impulse response. In this context the term "smoothing" is to be understood very broadly, i.e. as referring to any operation for smoothing or averaging the impulse response over the analysis period.

A first example of smoothing entails calculating the average of the matrix XX^h over the analysis period, which is assumed to include $\underline{m}$ training sequences:

$$L(XX^h) = \frac{1}{m}\sum_{1}^{m} XX^h$$

The operator $.^h$ represents the Hermitian transformation or complex conjugate transpose.

A second example of smoothing entails, after the ith training sequence is received, updating the smoothing matrix $L_{i-1}(XX^h)$ obtained on the (i-1)th training sequence by means of multiplier coefficient α, this factor generally being known as the smoothing forget factor and ranging from 0 to 1:

$$L_i(XX^h) = \alpha X_i X_i^h + (1-\alpha) L_{i-1}(XX^h)$$

Any means can be used for this initialization, including the first estimate X obtained, or an average obtained as described for a small number of training sequences.

For simplicity, the smoothing matrix $L(XX^h)$, which is in fact a statistical characterization matrix, is denoted L below.

The method then includes a step of looking for (eigenvalue, eigenvector) pairs of the characterization matrix.

This step is not described in more detail because it is well-known to the skilled person.

The eigenvalues λ_i are then classified in decreasing order. Their sum corresponds to the energy of the characterization signal X made up in part of a wanted signal which images the source signal and in part of the additive noise N of the transmission channel.

The dominant eigenvalues, the ones which are the

highest, represent the wanted signal, whereas the lowest eigenvalues represent noise.

A first option is to retain a predetermined number of dominant eigenvalues. For example, the first two eigenvalues λ_1 and λ_2 are retained for an impulse response with five coefficients.

A second option is to consider the wanted signal to have an energy which is a predetermined fraction <u>f</u> of the energy of the characterization signal. Thus using the notation λ_i for the eigenvalues with <u>i</u> varying from 1 to <u>p</u>, there will be <u>d</u> dominant eigenvalues, where <u>d</u> is obtained as follows:

$$\frac{\sum_{i=1}^{d} \lambda_i}{\sum_{i=1}^{p} \lambda_i} \leq f \quad \text{and} \quad \frac{\sum_{i=1}^{d+1} \lambda_i}{\sum_{i=1}^{p} \lambda_i} > f$$

The fraction <u>f</u> can be fixed *a priori*, for example at 95%. This fraction can also be derived from the signal-to-noise ratio of the receive signal obtained elsewhere.

A third option, and doubtless that offering the best performance, is for the additive noise N to be estimated directly from the receive signal and from the measurement matrix A. If N_0 denotes the noise vector affecting the receive signal:

$$S = AX + N_0$$

Given that the vectors S and N_0 have 22 components, the additive noise N can be expressed as follows:

$$N = \left(\frac{1}{22}\right)(S - AX)^h (S - AX)$$

This estimate of the additive noise can naturally be averaged and smoothed.

Using the previous notation, and standardizing the energies:

$$\sum_{i=d}^{p} \lambda_i \geq N \quad \text{and} \quad \sum_{i=d+1}^{p} \lambda_i < N$$

Thus the dominant eigenvalues are obtained from a direct estimate of the noise.

Whatever option is previously adopted, the next step of the method consists of calculating the correlation function of the source signal with the sum of the eigenvectors v_i associated with the dominant eigenvalues λ_i.

The source signal is oversampled relative to the bit period and is therefore denoted g(t) where t represents time and is a discrete variable whose quantization increment is 1/32 bit period, for example. It is represented by a vector with the same number of dimensions as the characterization signal, i.e. five dimensions in this example. The correction function c(t) is calculated for t varying from -1 to +1 bit period, for example, using the following expression:

$$c(t) = \sum_{i=1}^{d} g(t).\, v_i$$

The period between the source signal g(t) and the eigenvector v_i represents a scalar product, in the conventional way.

The final step of the method looks for the value t_0 of t that is closest to zero, which corresponds to the first relative maximum of the correlation c(t). It is this particular value t_0 which gives the required synchronization error relative to the reception signal.

The following complementary function c'(t) can also be considered:

$$c'(t) = \sum_{i=d+1}^{p} g(t).\, v_i$$

Note that the particular value t_0 mentioned above can also be obtained by seeking the value of t closest to

zero, which corresponds to the first relative minimum of the complementary formation c'(t).

These two methods of obtaining the synchronization t_0 are therefore equivalent.

Turning to Figure 2, a second embodiment of the invention is considered in which the characterization signal is the receive signal S, so that the source signal is now the reference signal, or, in the case of the GSM, the training sequence TS when it is GMSK (Gaussian minimum shift keying) modulated.

The statistic of the characterization signal is therefore estimated by means of a characterization matrix which is now obtained by smoothing the various occurrences of the receive signal S. Once again, the term "smoothing" is to be understood in a very broad sense in this context.

The characterization matrix L therefore takes the following form:

$$L(SS^h) = \frac{1}{m}\sum_{1}^{m} SS^h$$

or

$$L_i(SS^h) = \alpha S_i S_i^h + (1-\alpha) L_{i-1}(SS^h)$$

The p' eigenvalues λ'_i of the matrix are then looked for, and as in the first embodiment the d' dominant eigenvalues are identified.

The correlation function f(t) of the modulated training sequence and the sum of the eigenvectors v'_i associated with the dominant eigenvalues λ'_i are now calculated.

Once again, the training sequence g'(t) is oversampled and is represented by a vector with 22 components. The correlation function therefore becomes:

$$f(t) = \sum_{i=1}^{d'} g'(t).v'_i$$

As before, a new complementary function f'(t) can be

defined:

$$f'(t) = \sum_{i=d'+1}^{p'} g'(t).v'_i$$

The method terminates in the same manner by seeking the first maximum of the correlation function f(t) or the first minimum of the complementary function f'(t).

The invention can therefore be implemented in various ways, the essential point being to have access to a source signal and of the result of transmitting it, i.e. the characterization signal.

CLAIMS

1. A method of fine synchronization to a receive signal (S) corresponding to a reference signal (TS) transmitted in a transmission channel, characterized in that it includes the following steps:

- selecting a source signal producing a characterization signal (X, S) after it has passed through said transmission channel,
- establishing a characterization matrix (L) for estimating the covariance of said characterization signal (X, S),
- identifying dominant eigenvalues which are the highest eigenvalues (λ_i, λ'_i) of the characterization matrix (L),
- calculating the correlation function (c(t), f(t)) of said source signal with the sum of the eigenvectors (v_i, v'_i) associated with said dominant eigenvalues, and
- searching for the first maximum of the correlation function (c(t), f(t)).

2. A method according to claim 1, characterized in that the number (d, d') of dominant eigenvalues (λ_i, λ'_i) is predetermined.

3. A method according to claim 1, characterized in that the ratio of the sum of said dominant eigenvalues to the sum of all the eigenvalues is greater than or equal to a predetermined number.

4. A method according to claim 1, further including a step of estimating the additive noise (N) in the transmission channel, characterized in that said dominant eigenvalues are such that their sum is less than or equal to the sum of all the eigenvalues less said additive noise (N).

5. A method according to claim 4, characterized in that the additive noise (N) is estimated by normalizing the

instantaneous noise (N_0) which is evaluated by means of said receive signal (S), said reference signal (TS) and an estimate of the impulse response (X) of the transmission channel.

6. A method according to claim 5, characterized in that the expression for the instantaneous noise (N_0) is

$$N_0 = S - A.X$$

where A denotes the transmission matrix associated with said reference signal (TS).

7. A method according to claim 6, characterized in that said additive noise (N) is also averaged.

8. A method according to any of claims 1 to 7, characterized in that said characterization matrix (L) is the result of a smoothing operation.

9. A method according to any preceding claim, characterized in that said characterization signal is an estimate of the impulse response (X) of the transmission channel.

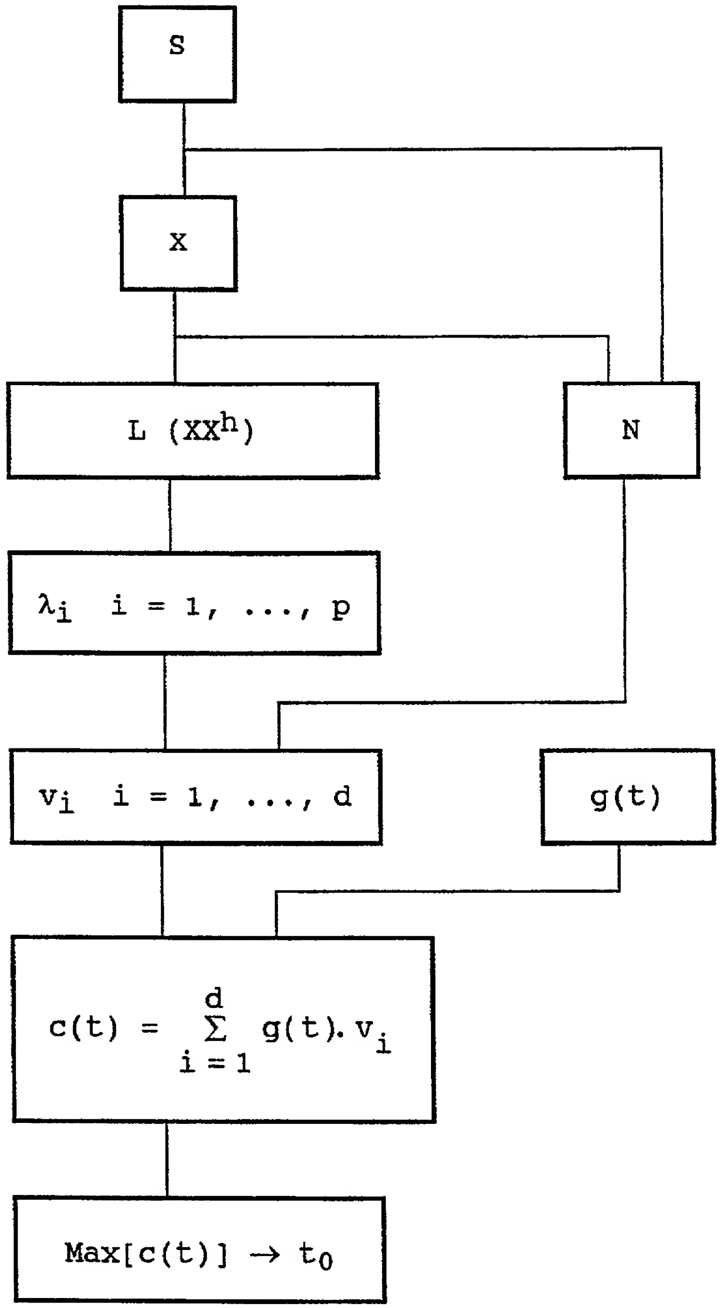
10. A method according to any of claims 1 to 8, characterized in that said characterization signal is said receive signal (S).

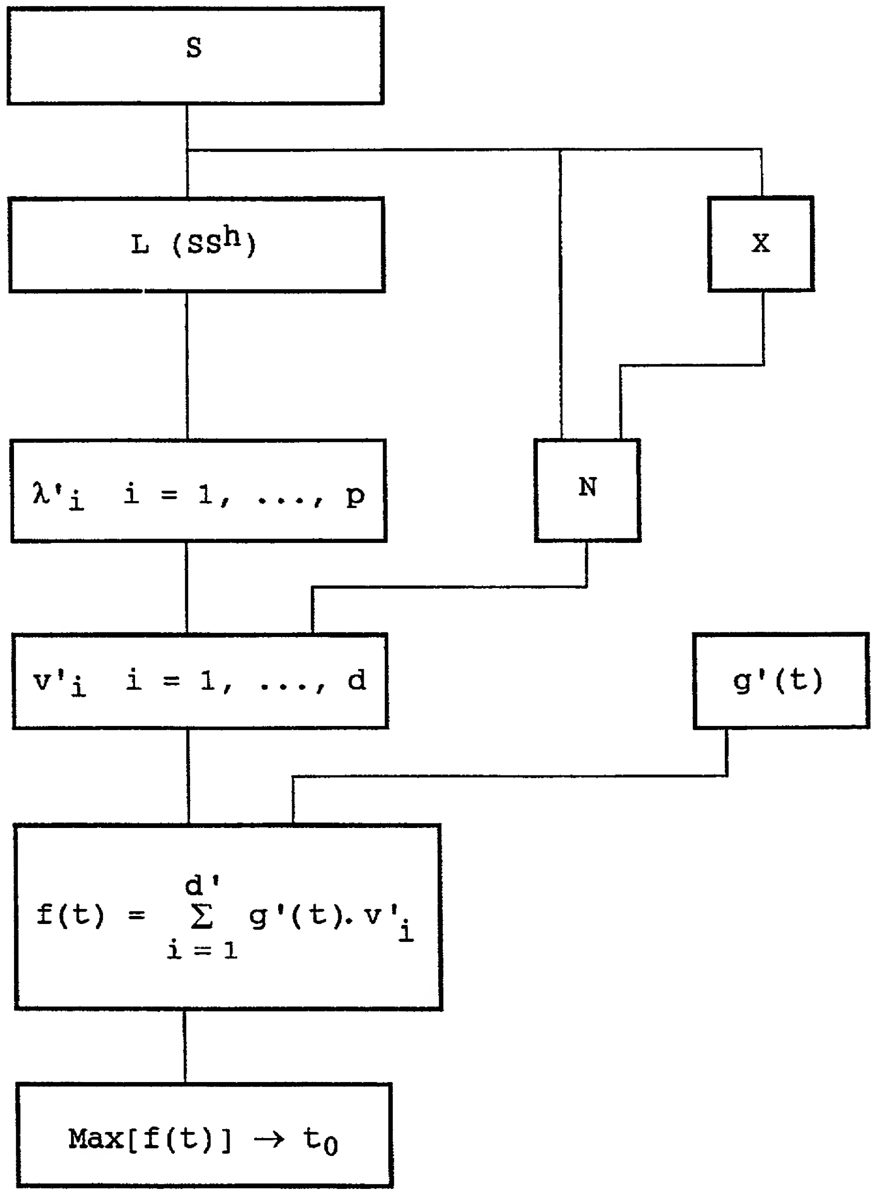
A B S T R A C T

METHOD OF FINE SYNCHRONIZATION TO A SIGNAL RECEIVED FROM A TRANSMISSION CHANNEL

The invention relates to a method of fine synchronization to a receive signal (S) corresponding to a reference signal (TS) transmitted in a transmission channel. The method includes the following steps:
- selecting a source signal producing a characterization signal (X) after it has passed through said transmission channel,
- establishing a characterization matrix (L) for estimating the covariance of the characterization signal (X),
- identifying dominant eigenvalues which are the highest eigenvalues (λ_i) of the characterization matrix (L),
- calculating the correlation function c(t) of the source signal with the sum of the eigenvectors (v_i) associated with the dominant eigenvalues, and
- searching for the first maximum of the correlation function c(t).

Translation of the title and the abstract as they were when originally filed by the Applicant. No account has been taken of any changes that may have been made subsequently by the PCT Authorities acting <u>ex officio</u>, e.g. under PCT Rules 37.2, 38.2, and/or 48.3.


S
X
L (XX^h)
N
λi i = 1, ..., p
vi i = 1, ..., d
g(t)
c(t) = Σ_{i=1}^{d} g(t). vi
Max[c(t)] → t0


S
L (ss^h)
X
λ'i i = 1, ..., p
N
v'i i = 1, ..., d
g'(t)
f(t) = Σ_{i=1}^{d'} g'(t).v'i
Max[f(t)] → t0

09/673002

DECLARATION AND POWER OF ATTORNEY FOR PATENT APPLICATION

As a below named inventor, I hereby declare that:

My residence, post office address and citizenship are as stated below next to my name.

I believe I am the original, first and sole inventor (if only one name is listed below) or an original, first and joint inventor (if plural names are listed below) of the subject matter which is claimed and for which a patent is sought on the invention entitled Procédé De Synchronisation fine Sur Un Signal Reçu D'un Canal de Transmission, the specification of which:

___ is attached hereto.

x was filed on April 9, 1999 as

Application Serial No. ________________

and was amended on __________ (if applicable).

I hereby state that I have reviewed and understand the contents of the above identified specification, including the claims, as amended by any amendment referred to above.

I acknowledge the duty to disclose information which is material to the examination of this application in accordance with Title 37, Code of Federal Regulations, Section 1.56(a).

I hereby claim foreign priority benefits under Title 35, United States Code, Section 119 of any foreign application(s) for patent or inventor's certificate listed below and have also identified below any foreign application for patent or inventor's certificate having a filing date before that of the application on which priority is claimed:

McWilliams, Sweeney & Ohlson, P.O. Box 2786, Chicago, Illinois 60690-2786, telephone number (312) 368-1300.

I hereby declare that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under Section 1001 of Title 18 of the United States Code and that such willful false statements may jeopardize the validity of the application or any patent issued thereon.

1-00

Full name of ~~sole or~~ first inventor: Jean-Louis DORNSTETTER

Signature ______ Date October 30th, 2000

Country of Residence: FRANCE JRX

Country of Citizenship: FRANCE

Post Office and Residence Address: 6 route de la grange aux Moines - 78460 Choisel

2-00

Full name of joint inventor: Nidham BEN RACHED

Signature ______ Date 30 oct. 2000

Country of Residence: FRANCE JRX

Country of Citizenship: TUNISIA

Post Office and Residence Address: 32, rue Baron
75017 Paris

Full name of joint inventor: ______

Signature ______ Date ______

Country of Residence: ______

Country of Citizenship: ______

Post Office and Residence Address: ______

PRIOR FOREIGN APPLICATION(S)

Country	Number	Date Filed	Priority Claimed Yes	No
FRANCE	98 04 782	April 10, 1998	X	

I hereby claim the benefit under Title 35, United States Code Section 120 of any United States application(s) listed below and, insofar as the subject matter of each of the claims of this application is not disclosed in the prior United States application in the manner provided by the first paragraph of Title 35, United States Code, Section 112, I acknowledge the duty to disclose material information as defined in Title 37, Code of Federal Regulations, Section 1.56(a) which occurred between the filing date of the prior application and the national or PCT international filing date of this application.

Application Serial No.	Filing Date	Status

And I hereby appoint Wm. Marshall Lee, Registration No. 16,853, John M. Mann, Registration No. 17,775, Thomas E. Smith, Registration No. 18,243, Dennis M. McWilliams, Registration No. 25,195, James R. Sweeney, Registration No. 18,721, William M. Lee, Jr., Registration No. 26,935, Glenn W. Ohlson, Registration No. 28,455, David C. Brezina, Registration No. 34,128, Jeffrey R. Gray, Registration No. 33,391, Timothy J. Engling, Registration No. 39,970, Gregory B. Beggs, Registration No. 19,286 and Gerald S. Geren, Registration No. 24,528 as my attorneys to prosecute this application and to transact all business in the Patent and Trademark Office connected herewith. It is requested that all communications be directed to Lee, Mann, Smith,